

# On the Log Determinant of Non-Central Wishart Matrices

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**Abstract** — We show that the expected log determinant of a complex non-central Wishart matrix is an increasing function of the non-centrality parameter. This demonstrates that the mutual information corresponding to an isotropically distributed Gaussian input to a multi-antenna Ricean fading channel is non-decreasing in the line-of-sight component.

We consider a memoryless Ricean fading channel with  $n$  transmitter antennas and  $m$  receiver antennas whose time- $k$  output  $\mathbf{Y}_k \in \mathcal{C}^m$  is given by

$$\mathbf{Y}_k = \mathbb{H}_k \mathbf{x}_k + \mathbf{Z}_k, \quad k = 1, 2, \dots, \quad (1)$$

where the vector  $\mathbf{x}_k \in \mathcal{C}^n$  denotes the channel input at time  $k$ , the additive noise process  $\{\mathbf{Z}_k\}$  is i.i.d.  $\sim \mathcal{N}_\mathcal{C}(0, N\mathbf{I}_m)$ , i.e., temporally and spatially white zero-mean variance- $N$  circularly symmetric complex Gaussian, and the channel matrix process  $\{\mathbb{H}_k\} = \{\mathbb{H}_k + \mathbf{D}\}$  is independent of  $\{\mathbf{Z}_k\}$  and temporally and spatially white unit-variance circularly symmetric complex Gaussian with the fading part  $\mathbb{H}_k$  i.i.d.  $\sim \mathcal{N}_\mathcal{C}(0, \mathbf{I}_m \otimes \mathbf{I}_n)$  and the deterministic line-of-sight gain part  $\mathbf{D} \in \mathcal{C}^{m \times n}$ . We assume that the receiver has perfect knowledge of the realization of  $\{\mathbb{H}_k\}$ . Under an isotropically distributed Gaussian input  $\mathbf{X} \sim \mathcal{N}_\mathcal{C}(0, \mathcal{E}_s \mathbf{I}_n)$ , the mutual information between the channel terminals is given by

$$I(\mathbf{X}; \mathbf{Y}, \mathbb{H}) = \mathbb{E} \left[ \log \det \left( \mathbf{I}_m + \frac{\mathcal{E}_s}{N} \mathbb{H} \mathbb{H}^\dagger \right) \right]. \quad (2)$$

Since  $\det(\mathbf{I}_m + \frac{\mathcal{E}_s}{N} \mathbb{H} \mathbb{H}^\dagger) = \det(\mathbf{I}_n + \frac{\mathcal{E}_s}{N} \mathbb{H}^\dagger \mathbb{H})$ , we assume without loss of generality that  $m \leq n$ , i.e., the number of receiver antennas does not exceed the number of transmitter antennas. Under this condition, the random matrix  $\mathbb{W} \triangleq \mathbb{H} \mathbb{H}^\dagger$  has the complex non-central Wishart distribution  $\mathcal{W}_n(m, \Lambda)$  [1], which is the matrix analogue of the non-central  $\chi^2$  distribution. In addition, since the distribution of the random matrix  $\mathbb{H}$  is rotationally symmetric, the mutual information (2) is a symmetric function of eigenvalues of  $\Lambda \triangleq \mathbf{D} \mathbf{D}^\dagger$ , and hence we assume that  $\Lambda$  is a nonnegative diagonal matrix with ordered diagonal entries  $\lambda_1 \geq \lambda_2 \geq \dots \geq \lambda_m$ .

While the isotropically distributed input  $\mathbf{X}$  need not achieve capacity, the corresponding mutual information (2) serves as a lower bound to the channel capacity, and is especially meaningful in the analysis of systems where the same codebook must be used under a variety of scattering environments. Expression such as (2) also arise in other scenarios, e.g., in the high SNR analysis of fading channel without receiver side information [2]. Note also that the mutual information (2) is indeed the capacity for a Rayleigh fading channel ( $\Lambda = 0$ ) [3].

<sup>1</sup>This research was conducted while Y. H. Kim was visiting the Swiss Federal Institute of Technology (ETH) Zurich.

Analytic or even numerical calculation of (2) is, however, intractable except for a few special cases. Instead we show that the mutual information (2) is an increasing function of the eigenvalues of  $\Lambda$ , providing a partial solution to an engineering question “Is a larger line-of-sight gain preferable?”.

**Theorem 1** Suppose that  $\mathbb{W}$  and  $\mathbb{W}'$  are complex non-central Wishart matrices with  $\mathbb{W} \sim \mathcal{W}_n(m, \Lambda)$  and  $\mathbb{W}' \sim \mathcal{W}_n(m, \Lambda')$  where  $\Lambda \succeq \Lambda'$  are nonnegative diagonal matrices with ordered diagonal entries. Then, for all  $\gamma \geq 0$ ,  $\det(\gamma \mathbf{I} + \mathbb{W})$  is stochastically larger than  $\det(\gamma \mathbf{I} + \mathbb{W}')$  and in particular,

$$\mathbb{E}[\log \det(\gamma \mathbf{I} + \mathbb{W})] \geq \mathbb{E}[\log \det(\gamma \mathbf{I} + \mathbb{W}')] .$$

The proof is based on the convexity of  $\det(\gamma \mathbf{I} + \mathbf{X} \mathbf{X}^\dagger)$  in each column of matrix  $\mathbf{X} \in \mathcal{C}^{m \times n}$  and the stochastic ordering for multivariate Gaussian distributions over a symmetric convex set [4].

From this theorem it is immediate that, in the presence of receiver-side channel information, the capacity of a Ricean fading channel is no less than that of a Rayleigh fading channel. Combining the Barlett decomposition for Wishart matrices and known results for the expected log determinant of the non-central  $\chi^2$  distribution [2], we obtain the following lower bound on the expected log determinant of a Wishart matrix and hence a lower bound on the capacity of a Ricean fading channel with channel state information at the receiver.

**Corollary 1** Let  $\mathbb{W} \sim \mathcal{W}_n(m, \Lambda)$ . Then

$$g_n(\lambda_1) + \sum_{k=2}^m \psi(n - k + 1) \leq \mathbb{E}[\log \det(\mathbb{W})] \leq \sum_{k=1}^m g_n(\lambda_k),$$

with equality in the lower bound if, and only if,  $\Lambda = \text{diag}(\lambda_1, 0, \dots, 0)$ . Here

$$g_n(\lambda) = \log(\lambda) - \text{Ei}(-\lambda) + \sum_{k=1}^{n-1} \left(\frac{-1}{\lambda}\right)^k \left[ e^{-\lambda} (k-1)! - \frac{(n-1)!}{k(n-1-k)!} \right],$$

$\psi$  is Euler's digamma function, and  $\text{Ei}$  is the exponential integral function.

The upper bound follows from Hadamard's inequality.

## REFERENCES

- [1] A. T. James, “Distributions of matrix variates and latent roots derived from normal samples,” *Ann. Math. Statist.*, vol. 35, no. 2, pp. 475–501, 1964.
- [2] A. Lapidoth and S. M. Moser, “Capacity bounds via duality with applications to multi-antenna systems on flat fading channels,” submitted to *IEEE Trans. Inform. Theory*, 2002.
- [3] I. E. Telatar, “Capacity of multi-antenna Gaussian channels,” *European Trans. Telecomm.*, vol. 10, no. 6, pp. 585–595, 1999.
- [4] T. W. Anderson, “The integral of a symmetric unimodal function over a symmetric convex set and some probability inequalities,” *Proc. Amer. Math. Soc.*, vol. 6, no. 2, pp. 170–176, 1955.